

# Linear Systems I

## Lecture 15

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## Review: stabilizability for LTI systems

$(A, B)$  uncontrollable:  $\text{rank } \mathcal{C} = \text{rank}[B \ AB \ A^2B \ \dots \ A^{n-1}B] = m < n$

$$\exists T \text{ (invertible)} : \begin{bmatrix} \dot{x}_c \\ \dot{x}_u \end{bmatrix} = \underbrace{\begin{bmatrix} A_c & A_{12} \\ 0 & A_u \end{bmatrix}}_{\bar{A}} \begin{bmatrix} x_c \\ x_u \end{bmatrix} + \underbrace{\begin{bmatrix} B_c \\ 0 \end{bmatrix}}_{\bar{B}} u$$

$$\bar{A} = T^{-1}AT, \quad \bar{B} = T^{-1}B$$

### Definition (Stabilizable LTI system)

**Def. (Stabilizable system):** The pair  $(A, B)$  is stabilizable if it is algebraically equivalent to a system in the standard form for uncontrollable systems with  $n = m$  (i.e.,  $A_u$  does not exist) or with  $A_u$  a stability matrix.

### Definition (Stabilizable LTI system (alternative definition))

The pair  $(A, B)$  is stabilizable if there exists a state feedback gain matrix  $K$  for which all the eigenvalues of  $A - BK$  have strictly negative real part.

There are various stabilizability tests. Following are some of them:

### Theorem

*The following statements are equivalent:*

- *The pair  $(A, B)$  is stabilizable;*
- *There exists no left eigenvector of  $A$  associated with an eigenvalue having nonnegative real part that is orthogonal to the columns of  $B$ ;*

$$\begin{cases} v^* A = \lambda v & (\operatorname{Re}[\lambda(A)] \geq 0) \\ v^* B = 0 \end{cases} \implies v = 0$$

- $\operatorname{rank}[\lambda I - A \quad B] = n$  for all  $\operatorname{Re}[\lambda(A)] \geq 0$ .

## Review: tests to check stabilizability of LTI systems: example

$$\dot{x} = \begin{bmatrix} -11 & 30 \\ -4 & 11 \end{bmatrix} x + \begin{bmatrix} 10 \\ 4 \end{bmatrix} u$$

Controllability text:

$$\text{rank } \mathcal{C} = \text{rank}[B \ AB] = \text{rank} \begin{bmatrix} 10 & 10 \\ 4 & 4 \end{bmatrix} = 1 \implies (A, B) \text{ is not controllable!}$$

### PBH eigenvalue test for controllability

- first find  $\lambda[A]$ :

$$\Delta(A) = \det(\lambda I - A) = (\lambda + 11)(\lambda - 11) + 120 = \lambda^2 - 1 = (\lambda - 1)(\lambda + 1) = 0 \implies \lambda[A] = \{-1, 1\}$$

- check rank of  $[\lambda I - A \ B]$  for  $\lambda[A] = \{-1, 1\}$

$$\lambda = -1: \text{rank}[-I - A \ B] = \text{rank} \begin{bmatrix} 10 & -30 & 10 \\ 4 & -12 & 4 \end{bmatrix} = 1 \implies \lambda = -1 \text{ is not a controllable eigenvalue}$$

$$\lambda = 1: \text{rank}[I - A \ B] = \text{rank} \begin{bmatrix} 12 & -30 & 10 \\ 4 & -10 & 4 \end{bmatrix} = 2 \implies \lambda = 1 \text{ is a controllable eigenvalue}$$

Then:

- $(A, B)$  is not controllable
- $(A, B)$  is stabilizable: because  $\text{rank}[\lambda I - A \ B] = n$  for the eigenvalue with positive real part  $\lambda = 1$ .

# Regulation via state-feedback control when (A,B) is controllable: pole-placement/eigenvalue placement

$$\dot{x} = Ax + Bu, \quad u = -Kx \Rightarrow \dot{x} = (A - BK)x$$

- (A, B) controllable: Given any symmetric set of  $n$  complex numbers  $\{\nu_1, \nu_2, \dots, \nu_n\}$ , there exists a full-state feedback matrix  $K$  such that the closed-loop system matrix  $(A - BK)$  has eigenvalues equal to these  $\nu_i$ 's.

$$\exists K: \det(\lambda I - (A - BK)) = \underbrace{(\lambda - \nu_1)(\lambda - \nu_2) \cdots (\lambda - \nu_n)}_{\text{desired charac. polynomial}}$$

- Single input systems ( $u \in \mathbb{R}$ ,  $B \in \mathbb{R}^{n \times 1}$ ): See HW 6 for a procedure for eigenvalue placement. You can also use Achermann formula.
- Multi-input systems ( $u \in \mathbb{R}^p$ ,  $B \in \mathbb{R}^{n \times p}$ ): Theorem below gives a solution

*Theorem: Suppose  $(A \in \mathbb{R}^{n \times n}, B \in \mathbb{R}^{n \times p})$  is controllable. Then  $(A + BF, Bv)$ , is controllable for almost any  $F \in \mathbb{R}^{p \times n}$  and  $v \in \mathbb{R}^{p \times 1}$ .*

**State feedback for multi input systems:** Example, place the eigenvalues at  $\{-1, -2, -3\}$

$$A = \begin{bmatrix} 1 & 0 & 0 \\ 1 & 2 & 0 \\ 4 & 5 & 1 \end{bmatrix}, \quad B = \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 2 & 0 \end{bmatrix}$$

- choose  $F \in \mathbb{R}^{p \times n}$  and  $v \in \mathbb{R}^{p \times 1}$  such that  $(A + BF, Bv)$  is controllable

$$F = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}, \quad v = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \Rightarrow \bar{A} = A + BF = \begin{bmatrix} 2 & 0 & 0 \\ 1 & 3 & 0 \\ 6 & 5 & 1 \end{bmatrix}, \quad \bar{B} = Bv = \begin{bmatrix} 1 \\ 0 \\ 2 \end{bmatrix}, \quad \text{rank}(\bar{C}) = \text{rank} \begin{bmatrix} 1 & 2 & 4 \\ 0 & 1 & 5 \\ 2 & 8 & 25 \end{bmatrix}$$

- place eigenvalues of  $(\bar{A}, \bar{B})$  at your desired location  $\{-1, -2, -3\}$  using the methods for single input systems: here I get  $\bar{K} = [28 \ 80 \ -8]$
- the  $K$  in  $A - BK$  is obtained from  $A - BK = A + BF - Bv\bar{K} = A - B \underbrace{(-F + v\bar{K})}_K$ , which gives

$$K = \begin{bmatrix} 27 & 80 & -8 \\ 0 & -1 & 0 \end{bmatrix}$$

- For this example  $A - BK = \begin{bmatrix} -26 & -80 & 8 \\ 1 & 3 & 0 \\ -50 & -155 & 17 \end{bmatrix}$ , with eigenvalues at the desired location  $\{-1, -2, -3\}$ .

## Regulation via state-feedback control when $(A, B)$ is stabilizable: pole-placement/eigenvalue placement

$$\dot{x} = Ax + Bu, \quad u = -Kx \Rightarrow \dot{x} = (A - BK)x$$

- $(A, B)$  is not controllable:  $\text{rank}(C) = m < n$  ( $A \in \mathbb{R}^{n \times n}$ ):

$$\exists T \text{ invertible} : x = T\bar{x} : \quad \dot{\bar{x}} = \begin{bmatrix} \dot{\bar{x}}_c \\ \dot{\bar{x}}_u \end{bmatrix} = \underbrace{\begin{bmatrix} A_c & A_{12} \\ 0 & A_u \end{bmatrix}}_{\bar{A} = T^{-1}AT} \begin{bmatrix} x_c \\ x_u \end{bmatrix} + \underbrace{\begin{bmatrix} B_c \\ 0 \end{bmatrix}}_{\bar{B} = T^{-1}B} u$$

$$u = -Kx = -KT\bar{x} = -\bar{K}\bar{x} = -\begin{bmatrix} \bar{K}_1 & \bar{K}_2 \end{bmatrix} \begin{bmatrix} x_c \\ x_u \end{bmatrix}$$

$$\dot{\bar{x}} = \begin{bmatrix} \dot{\bar{x}}_c \\ \dot{\bar{x}}_u \end{bmatrix} = \underbrace{\begin{bmatrix} A_c - B_c \bar{K}_1 & A_{12} - B_c \bar{K}_2 \\ 0 & A_u \end{bmatrix}}_{\bar{A} - \bar{B} \bar{K} = T^{-1}(A - BK)T} \begin{bmatrix} x_c \\ x_u \end{bmatrix}$$

$(A_c, B_c)$  is controllable, we can place eigenvalues of  $(A_c - B_c \bar{K}_1)$  in any location we want using state feedback!

We can only change the location of controllable eigenvalues using state feedback

We can only stabilize a system whose uncontrollable eigenvalues are stable

- Transfer  $(A, B)$  to the controllable decomposition form
- Recall that  $\text{eig}(A) = \text{eig}(A_c) \cup \text{eig}(A_u)$
- $(A_c, B_c)$  is controllable, so you can place the eigenvalues of this controllable part at your desired locations using gain  $\bar{k}_1$ , i.e., eigenvalues of  $A_c - B_c \bar{k}_1$
- you can find the gain  $K$  placing the controllable eigenvalues in your desired places using  $K = \bar{K}T^{-1}$ , with  $\bar{K} = [\bar{k}_1 \quad \bar{k}_2]$ . You can set  $k_2$  to zero.
- you should arrive at  $\text{eig}(A - BK) = \text{eig}(A_c - B_c \bar{k}_1) \cup \text{eig}(A_u)$

## State feedback design for a stabilizable system: example

$$\dot{x} = \begin{bmatrix} -11 & 30 \\ -4 & 11 \end{bmatrix} x + \begin{bmatrix} 10 \\ 4 \end{bmatrix} u$$

Controllability text:

$$\text{rank } \mathcal{C} = \text{rank}[B \quad AB] = \text{rank} \begin{bmatrix} 10 & 10 \\ 4 & 4 \end{bmatrix} = 1 \implies (A, B) \text{ is not controllable!}$$

$$\text{eig}(A): \Delta(A) = \det(\lambda I - A) = (\lambda + 11)(\lambda - 11) + 120 = \lambda^2 - 1 = (\lambda - 1)(\lambda + 1) = 0 \Rightarrow \text{eig}(A) = \{-1, 1\}$$

Controllable decomposition

$$T = \begin{bmatrix} 5 & 0 \\ 2 & 0.2 \end{bmatrix}, \quad T^{-1} = \begin{bmatrix} 0.2 & 0 \\ -2 & 5 \end{bmatrix}$$
$$\bar{A} = T^{-1}AT = \begin{bmatrix} 0.2 & 0 \\ -2 & 5 \end{bmatrix} \begin{bmatrix} -11 & 30 \\ -4 & 11 \end{bmatrix} \begin{bmatrix} 5 & 0 \\ 2 & 0.2 \end{bmatrix} = \begin{bmatrix} 1 & -1.2 \\ 0 & -1 \end{bmatrix}, \quad \bar{B} = T^{-1}B = \begin{bmatrix} 2 \\ 0 \end{bmatrix}$$

**Objective** Place eigenvalues of  $A - BK$  at  $\{-1, -3\}$

We use  $(A_c, B_c) = (1, 2)$  to place eigenvalue of the controllable part at  $-3$ :  $\lambda(A_c - B_c \bar{k}_1) = -3$

$$\lambda - (1 - 2\bar{k}_1) = \lambda + 3 \Rightarrow \bar{k}_1 = 2.$$

$$u = \bar{K}\bar{x} = [2 \quad 0] \bar{x} = [2 \quad 0] T^{-1}x = [2 \quad 0] \begin{bmatrix} 0.2 & 0 \\ -2 & 5 \end{bmatrix} x = \underbrace{[0.4 \quad 0]}_K x.$$

You can confirm this by checking eigenvalues of  $A - BK = \begin{bmatrix} -11 & 30 \\ -4 & 11 \end{bmatrix} - \begin{bmatrix} 10 \\ 4 \end{bmatrix} [0.4 \quad 0] = \begin{bmatrix} -15 & 30 \\ -5.6 & 11 \end{bmatrix}$ .

## State feedback design for a stabilizable system: example (alternative approach)

Consider the state feedback  $u = -Kx = -[k_1 \ k_2]x$

$$A_{cl} = A - BK = \begin{bmatrix} -11 - 10k_1 & 30 - 10k_2 \\ -4 - 4k_1 & 11 - 4k_2 \end{bmatrix}$$

$$\Delta(A_{cl}) = \Delta(A - BK) = \det \left( \lambda I - \begin{bmatrix} -11 - 10k_1 & 30 - 10k_2 \\ -4 - 4k_1 & 11 - 4k_2 \end{bmatrix} \right) = (\lambda + 1)(\lambda + 10k_1 + 4k_2 - 1) = 0$$

$$\text{eig}(A_{cl}) = \{-1, -10k_1 - 4k_2 + 1\}$$

- Notice that we cannot change the location of uncontrollable eigenvalue but we can put the controllable eigenvalue in any new location using state feedback!
- We can pick  $k_1$  and  $k_2$  such that  $A_{cl}$  has eigenvalues with strictly negative real parts and, as such, stabilize the closed-loop system using  $u = -Kx$ .
- For example  $k_1 = 0$  and  $k_2 = 1$  results in  $\lambda[A_{cl}] = \{-1, -3\}$ .

You can confirm this by checking eigenvalues of

$$A - BK = \begin{bmatrix} -11 & 30 \\ -4 & 11 \end{bmatrix} - \begin{bmatrix} 10 \\ 4 \end{bmatrix} \begin{bmatrix} 0 & 1 \end{bmatrix} = \begin{bmatrix} -11 & 20 \\ -4 & 7 \end{bmatrix}.$$

Next Lecture

$$\begin{cases} \dot{x} = Ax + Bu, & x \in \mathbb{R}^n, u \in \mathbb{R}^p \\ y = Cx + Du, & y \in \mathbb{R}^q \end{cases} \quad x(0) = x_0 \in \mathbb{R}^n \quad (\star)$$

**Question of interest in Observability:** Can we reconstruct  $x(0)$  by knowing  $y(t)$  and  $u(t)$  over some finite time interval  $[0, t_1]$ ? (By knowing the initial condition, we can reconstruct the entire state  $x(t)$ , then use it in our state feedback to control the system)

$$y(t) = Ce^{At}x(0) + C \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau + Du(t)$$

$$\begin{cases} \dot{x} = Ax + Bu, & x \in \mathbb{R}^n, u \in \mathbb{R}^p \\ y = Cx + Du, & y \in \mathbb{R}^q \end{cases} \quad x(0) = x_0 \in \mathbb{R}^n \quad (\star)$$

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$$y(t) = Ce^{At}x(0) + C \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau + Du(t) \Leftrightarrow \bar{y}(t) = Ce^{At}x(0)$$

$$\bar{y}(t) = y(t) - C \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau - Du(t)$$

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The LTI state-space equation  $(\star)$  is said **to be observable** if for any unknown initial state  $x(0)$ ,  $\exists$  finite time  $t_1 > 0$  such that the knowledge of the input  $u$  and the output  $y$  over  $[0, t_1]$  suffices to determine uniquely the initial state  $x(0)$ . Otherwise, the equation is said to be unobservable.

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# Tests for Observability of LTI systems

The following statements are equivalent:

1 the  $n$ -dimensional pair  $(A, C)$  is observable

2 The  $n \times n$  matrix  $W_o(t) = \int_0^t e^{A^T \tau} C^T C e^{A \tau} d\tau$  is nonsingular for all  $t > 0$ .

3 Let  $\mathcal{O} = \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix}_{nq \times n}$  be the observability matrix, then  $\text{rank}(\mathcal{O}) = n$

4  $\text{rank} \begin{bmatrix} \lambda I - A \\ C \end{bmatrix} = n$  for all complex  $\lambda$

5  $\text{rank} \begin{bmatrix} \lambda I - A \\ C \end{bmatrix} = n$  for all  $\lambda$  eigenvalues of  $A$

6 If in addition, all eigenvalues of  $A$  have negative real parts, then the unique solution of

$$A^T W_o + W_o A = -C^T C$$

is positive definite. The solution is called the observability Gramian and can be expressed as

$$W_o = \int_0^{\infty} e^{A^T \tau} C^T C e^{A \tau} d\tau$$

## Review of controllable decomposition

$$\begin{aligned}\dot{x} &= Ax + Bu, & x \in \mathbb{R}^n, & u \in \mathbb{R}^p \\ y &= Cx + Du, & y \in \mathbb{R}^q\end{aligned}$$

### Theorem

$$\text{rank} [B \quad AB \quad \dots \quad A^{n-1}B] = m < n$$

$\exists T$  invertible s.t.  $\bar{x} = T^{-1}x$  transforms state equations to

$$\bar{A} = T^{-1}AT = \begin{bmatrix} A_c & A_{12} \\ 0 & A_u \end{bmatrix}, \quad \bar{B} = T^{-1}B = \begin{bmatrix} B_c \\ 0 \end{bmatrix}$$

$$\bar{C} = [C_u \quad C_c], \quad \bar{D} = D,$$

$$A_c \in \mathbb{R}^{m \times m}, \quad B_c \in \mathbb{R}^{m \times p}, \quad C_c \in \mathbb{R}^{q \times m},$$

$$T = \left[ \underbrace{t_1 \quad t_2 \quad \dots \quad t_m}_{\substack{\text{\color{blue} }m \text{ linearly independent} \\ \text{\color{blue} } \text{columns of } C}} \quad \left| \quad \underbrace{t_{m+1} \quad t_{m+2} \quad \dots \quad t_n}_{\substack{\text{any way you can} \\ \text{s.t. all columns of} \\ T \text{ are linearly independent}}} \right. \right]$$

$(A_c, B_c)$  is controllable!

$$G(s) = \bar{G}(s) = \bar{C}(sI - \bar{A})^{-1}\bar{B} + \bar{D} = C_c(sI - A_c)^{-1}B_c + D$$

# Observable decomposition

$$\begin{aligned}\dot{x} &= Ax + Bu, & x \in \mathbb{R}^n, & u \in \mathbb{R}^p \\ y &= Cx + Du, & y \in \mathbb{R}^q\end{aligned}$$

## Theorem

$\exists T$  invertible s.t.  $\bar{x} = T^{-1}x$  transforms state equations to

$$\text{rank} \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix} = \bar{m} < n :$$

$$\bar{A} = T^{-1}AT = \begin{bmatrix} A_o & 0 \\ A_{12} & A_{\bar{o}} \end{bmatrix}, \quad \bar{B} = T^{-1}B = \begin{bmatrix} B_o \\ B_{\bar{o}} \end{bmatrix}$$

$$\bar{C} = CT = [C_o \quad 0], \quad \bar{D} = D,$$

$$A_o \in \mathbb{R}^{\bar{m} \times \bar{m}}, \quad B_o \in \mathbb{R}^{\bar{m} \times p}, \quad C_o \in \mathbb{R}^{q \times \bar{m}},$$

$$T = \left[ \begin{array}{c|c} \underbrace{t_1 \quad t_2 \quad \cdots \quad t_{\bar{m}}}_{\substack{\text{any way you can} \\ \text{s.t. all columns of} \\ T \text{ are linearly independent}}} & \underbrace{t_{\bar{m}+1} \quad t_{\bar{m}+2} \quad \cdots \quad t_n}_{\substack{n - \bar{m} \text{ linearly independent} \\ \text{vectors spanning the} \\ \text{nullspace of } \mathcal{O}}} \end{array} \right]$$

$(A_o, B_o)$  is observable.

$$G(s) = \bar{G}(s) = \bar{C}(sI - \bar{A})^{-1}\bar{B} + \bar{D} = C_o(sI - A_o)^{-1}B_o + D$$

# Kalman decomposition

$$\dot{x} = Ax + Bu, \quad y = Cx + Du, \quad x \in \mathbb{R}^n, \quad u \in \mathbb{R}^p, y \in \mathbb{R}^q$$

## Theorem

$\exists T$  invertible s.t.  $\bar{x} = T^{-1}x$  transforms state equations to

$$\text{rank} \begin{bmatrix} B & AB & \dots & A^{n-1}B \end{bmatrix} = m < n$$

$$\text{rank} \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix} = \bar{m} < n$$

$$\begin{bmatrix} \dot{x}_{c0} \\ \dot{x}_{c\bar{o}} \\ \dot{x}_{\bar{e}o} \\ \dot{x}_{\bar{e}\bar{o}} \end{bmatrix} = \underbrace{\begin{bmatrix} A_{c0} & 0 & x_o & 0 \\ A_{cx} & A_{c\bar{o}} & A_{xx} & A_{x\bar{o}} \\ 0 & 0 & A_{\bar{e}o} & 0 \\ 0 & 0 & A_{\bar{e}x} & A_{\bar{e}\bar{o}} \end{bmatrix}}_{\bar{A} = T^{-1}AT} \begin{bmatrix} x_{c0} \\ x_{c\bar{o}} \\ x_{\bar{e}o} \\ x_{\bar{e}\bar{o}} \end{bmatrix} + \underbrace{\begin{bmatrix} B_{c0} \\ B_{c\bar{o}} \\ 0 \\ 0 \end{bmatrix}}_{\bar{B} = T^{-1}B} u$$

$$y = \underbrace{\begin{bmatrix} C_{c0} & 0 & C_{\bar{e}o} & 0 \end{bmatrix}}_{\bar{C} = CT} \begin{bmatrix} x_{c0} \\ x_{c\bar{o}} \\ x_{\bar{e}o} \\ x_{\bar{e}\bar{o}} \end{bmatrix} + Du,$$

$$T = \begin{bmatrix} T_{c0} & T_{c\bar{o}} & T_{\bar{e}o} & T_{\bar{e}\bar{o}} \end{bmatrix}$$

- columns of  $[T_{c0} \ T_{c\bar{o}}]$  span the  $\text{Im}C$
- columns of  $T_{c\bar{o}}$  span the  $\text{null}O \cap \text{Im}C$
- columns of  $[T_{c\bar{o}} \ T_{\bar{e}\bar{o}}]$  span the  $\text{null}O$
- columns of  $T_{\bar{e}o}$  along with the elements described above construct an invertible  $T$
- $(A_{c0}, B_{c0}, C_{c0})$  is both controllable and observable.
- $\left( \begin{bmatrix} A_{c0} & 0 \\ A_{cx} & A_{c\bar{o}} \end{bmatrix}, \begin{bmatrix} B_{c0} \\ B_{c\bar{o}} \end{bmatrix} \right)$  is controllable
- $\left( \begin{bmatrix} A_{c0} & A_{xo} \\ 0 & A_{\bar{e}o} \end{bmatrix}, [C_{c0} \ C_{\bar{e}o}] \right)$  is controllable

$$G(s) = \bar{G}(s) = \bar{C}(sI - \bar{A})^{-1}\bar{B} + \bar{D} = C_{c0}(sI - A_{c0})^{-1}B_{c0} + D$$